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A hybrid dynamical system approach to the impulsive control of spacecraft rendezvous

Alexandre Seuret 1, Rafael Vazquez 1 and Luca Zaccarian2

Abstract

This paper introduces a hybrid dynamical system methodology for managing impulsive control in spacecraft rendezvous and proximity operations under the Hill-Clohessy-Wiltshire model. We address the control design problem by isolating the out-of-plane from the in-plane dynamics and present a feedback control law for each of them. This law is based on a Lyapunov function tailored to each of the dynamics, capable of addressing thruster saturation and also a minimum impulse bit. These Lyapunov functions were found by reformulating the system's dynamics into coordinates that more intuitively represent their physical behavior. The effectiveness of our control laws is then shown through numerical simulation.

I. Introduction

From the increasing complexity of space missions, emerged the needs of servicing satellites effectively. Operations such as inspection, repair, refueling, and monitoring are essential, requiring a spacecraft, known as the chaser, to execute precise maneuvers near a target spacecraft. These maneuvers, known as rendezvous and proximity operations, are critical in guiding a spacecraft to a pre-determined proximity to the target to perform mission-specific tasks. The demand for autonomous guidance and control in these tasks is more pressing than ever, motivated by key space activities like asteroid mining [1], collision avoidance [2], on-orbit assembly [3], debris removal [4], and resupplying missions [5]. Originating from the Apollo program's lunar orbital rendezvous concept [6], which was essential for reducing payload mass and the feasibility of the mission, the techniques have considerably developed. Proximity operations have since become commonplace, such as in the frequent rendezvous missions to the International Space Station, and continue to be integral in the advancement of low Earth orbit operations and beyond. The most basic rendezvous model is described by the Clohessy-Wiltshire (HCW) equations [7], which were in fact developed for the Apollo program and consider the target in a circular Keplerian orbit and the chaser in close proximity.

In the past, many works have addressed the design of feedback laws for the rendezvous problem by using a Model Predictive Control (MPC) strategy, see e.g. [8], [9], [10], [11]. MPC can contribute important properties such as optimality and constraint handling, however at a considerable computational cost that may not be possible e.g. onboard low-cost satellites such as Cubesats; in addition it is not a trivial task to provide guarantees on stability and feasibility with MPC, or to handle minimum impulse bits. On the other hand, despite the potential of hybrid systems theory [12] for ensuring stability, simplifying calculations, and improving efficiency, it has rarely been applied to the rendezvous problem, with limited exceptions such as [13] (and references therein). Our approach diverges from these instances by utilizing a simpler simulation model but incorporating saturations into our analysis, and avoiding the need for an optimization process in the control law computation.

This paper considers the terminal rendezvous stage with a focus on designing efficient impulsive maneuvers. Traditional planning for such missions simplifies the process by approximating actual maneuvers with instantaneous velocity changes, a widely accepted practice that justifies the impulsive approach we use in this paper. Within a hybrid systems framework [12], we separate the out-of-plane and in-plane dynamics and propose a feedback control law based on a specific Lyapunov functions for each of the dynamics that is able to both thruster saturation and minimum impulse bit. These Lyapunov functions are found by expressing the dynamics in more natural coordinates that capture their physical behaviour. Finally, our control law's performance is validated through simulations.

The structure of the manuscript is as follows: Section II introduces the HCW model used in our approach. Section III outlines the hybrid stabilization for the out-of-plane dynamics whereas Section IV similarly deals with the stabilization of the in-plane dynamics. The different stabilizers are combined in a unique feedback law whose properties are established in Section V. Next, Section VI provides simulations results, and the paper is closed in Section VII with some concluding remarks. Due to space limitation, the proofs have been removed but can be found in [14].

II. PROBLEM STATEMENT

There are numerous mathematical models for spacecraft rendezvous. if the target is orbiting in a *circular* Keplerian orbit and the approaching vehicle (chaser) is close to the target, then the linear Hill-Clohessy-Wiltshire (HCW) equations, introduced in [15] and [7], describe with adequate precision the relative position of the spacecraft.

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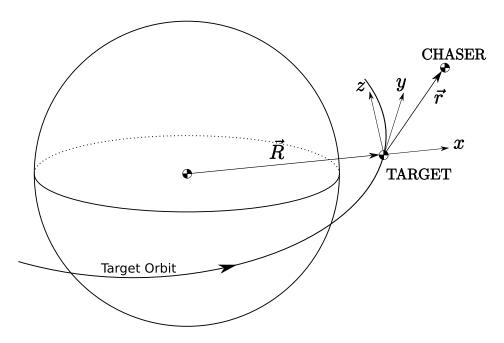


Fig. 1. Local-Vertical, Local-Horizontal (LVLH) frame.

A. HCW model

The HCW model assumes that the target vehicle is passive and moving along a circular orbit. Using impulsive control, it describes the relative motion of a chaser vehicle close to the target and can be formulated as the following equation.

$$\ddot{r}_x - 3n^2 r_x - 2n\dot{r}_y = 0, (1a)$$

$$\ddot{r}_y + 2n\dot{r}_x = 0, (1b)$$

$$\ddot{r}_z + n^2 r_z = 0, (1c)$$

where

- $r = (r_x, r_y, r_z) \in \mathbb{R}^3$ stands for the relative position between chaser and target in the target reference frame.
- To avoid any confusion, the time derivative of (r_x, r_y, r_z) will be denoted as $v = (v_x, v_y, v_z)$, which stands for the relative velocities between the chaser and the target in the target reference frame.
- n is the mean orbital angular speed of the target, which, in this case being the target's orbit circular, coincides with its instantaneous angular speed. The angular speed of the target through its orbit is $n = \sqrt{\frac{\mu}{R^3}}$, where μ is the gravitation parameter of the Earth, $\mu = 398600.4 \text{ km}^3/\text{s}^2$ and R is the radius. Thus for a typical orbit at, say, an altitude of 500 kilometers we would get n = 0.0011 rad/s.

Thus the full state is characterized by $(r, v) \in \mathbb{R}^6$. Equations (1) are expressed in a local target reference frame (LVLH), which is a rotating frame centered at the target (see Figure 1). The coordinates lend themselves to a physical explanation: r_x is the radial distance (with a positive r_x , the chaser is at a higher altitude than the target, with a negative r_x it is at a lower altitude); r_y is the along-track distance (the phase with respect to the target's orbit). Thus, $r_x = r_z = 0$ and $r_y \neq 0$ are equilibria, representing the chaser in the same orbit as the target but with some phase lag: delayed with respect to it or ahead with respect to it. The $r_x - r_y$ motion is coupled due to the laws of orbital mechanics.

Finally, r_z is the relative vertical distance to the target's orbit, describing how the plane of the chaser's orbit is related to the orbital plane of the target. Physically, the r_z motion is decoupled from the other ones because orbital mechanics dictates that orbital motions stay in a plane; under the HCW approximation, nothing that one can do in the $r_x - r_y$ plane can change this plane, thus r_z remains decoupled, and vice-versa: changing the plane does not affect what happens in $r_x - r_y$. There is no drift in r_z as seen in the equations, since the chaser's orbital plane remains the same if one does not act on this direction.

B. Propulsive model and constraints

In this paper, we consider that the control action is performed as impulses acting only on the velocities, that is, at specific instants decided by the control law, the velocities experience the following discontinuous motion

$$v_x^+ = v_x + u_x,\tag{2a}$$

$$v_y^+ = v_y + u_y,$$
 (2b)
 $v_z^+ = v_z + u_z,$ (2c)

$$v_z^+ = v_z + u_z, \tag{2c}$$

where we defined the control input $u = (u_x, u_y, u_z) \in \mathbb{R}^3$.

Model (1),(2) shows that the dynamics of a spacecraft rendezvous are governed by open-loop continuous-time dynamics. The only control action occurs at some time instants. It consists in an abrupt change of the velocity in all the directions of the target reference frame. Regarding the possible values of thrust (namely of u), we assume that the control inputs are bounded in absolute value. This means that there exists $u_{\rm M}>0$ such that

$$|u_x| \le u_{\mathcal{M}}, \quad |u_y| \le u_{\mathcal{M}}, \quad |u_z| \le u_{\mathcal{M}}, \tag{3}$$

We assume that $|u_{(.)}|$ can take any value in the interval, i.e., it is assumed that thrusters valves can be opened partially to produce the exact amount of force commanded by the control law. Thus, we define the symmetric saturation nonlinear map sat, whose components are defined as follows

$$\operatorname{sat}_{i}(u_{i}) = \begin{cases} \operatorname{sign}(u_{i})u_{M}, & \text{if } |u_{i}| > u_{M}, \\ u_{i}, & \text{if } 0 \leq |u_{i}| \leq u_{M}. \end{cases}$$

$$(4)$$

Remark 1: In practice, the thrusters may also be bounded from below, i.e. control inputs that are too small cannot be produced by the thrusters. This issue will not be considered in this paper and is left as future work.

C. Control objectives

In view of the hybrid nature of the spacecraft rendezvous with mixed continuous and discrete dynamics, we propose here to use the hybrid dynamical systems theory [16], [12], to propose nonlinear hybrid control laws for system (1)-(2) accounting for the input constraints (4). The idea to deploy the Hybrid Dynamical Systems framework on spacecraft rendezvous is not new since it has already been adopted in previous research [13], [17]. The main difference in this paper is that we will treat the rendezvous problem directly using the HCW model. More precisely, we will present the following contributions:

- a hybrid control law for u_z to stabilize the out-of-plane dynamics (the z direction);
- two independent hybrid control laws for u_x and u_y to stabilize the in-plane dynamics (the x, y plane);
- each control law aims at defining both the correct position to perform efficient actions by the thrusters and the appropriate magnitude of the impulses;
- ullet the control impulses u_x , u_y and u_z are not necessarily synchronized, to allow for more flexibility and to potentially reduce the number of impulses;
- the nonlinear control laws are equipped with a dwell-time dynamics to avoid fast consecutive use of the same input u_x , u_y , or u_z ; this dwell-time property, among other things, avoids the Zeno phenomenon i.e., the occurrence of an infinite number of impulses in a finite interval of time.

In the HCW model (1a)–(1b), one can identify two decoupled dynamics, referred to as in-plane (the (x, y) directions) and out-of-plane dynamics (the z direction). We carry out the hybrid stabilizers by first focusing on the z (out-of-plane) direction and then show how the ensuing ideas can be followed for stabilizing also the more challenging in-plane direction.

D. Hybrid Dynamical Systems framework

The chaser's maneuver is represented as a sudden velocity change, given by u. As such, the governing dynamics are considered a hybrid system, which integrates both continuous and discrete dynamical characteristics, permitting time-evolving processes and instantaneous transitions. Following the definition of hybrid dynamical systems from [16],[12], such a system can be formally defined as:

$$\mathcal{H} = (\mathcal{C}, \mathcal{D}, F, G) : \begin{cases} \dot{\mathbf{x}} \in F(\mathbf{x}), & \mathbf{x} \in \mathcal{C} \\ \mathbf{x}^+ \in G(\mathbf{x}), & \mathbf{x} \in \mathcal{D} \end{cases}$$
(5a)

where $n_x \in \mathbb{N}$ is the state dimension, $\mathcal{C} \subseteq \mathbb{R}^{n_x}$ ($\mathcal{D} \subseteq \mathbb{R}^{n_x}$) is the flow (jump) set, $F : \mathcal{C} \to \mathbb{R}^{n_x}$ is the flow map and $G: \mathcal{D} \to \mathbb{R}^{n_x}$ the jump map. Mathematically, \mathcal{H} is given by

In the next section, we will use this framework to develop hybrid control laws for the spacecraft rendezvous problem.

III. STABILIZATION IN THE z DIRECTION

When focusing on the z direction, one can extract the following essential dynamics from the model (1c) and (2c):

$$\begin{bmatrix} \dot{r}_z \\ \dot{v}_z \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -n^2 & 0 \end{bmatrix} \begin{bmatrix} r_z \\ v_z \end{bmatrix} \quad \text{when in free motion,}$$

$$\begin{bmatrix} r_z^+ \\ v_z^+ \end{bmatrix} = \begin{bmatrix} r_z \\ v_z \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u_z \quad \text{when firing the input } u_z.$$

$$(6)$$

$$\begin{bmatrix} r_z^+ \\ v_z^+ \end{bmatrix} = \begin{bmatrix} r_z \\ v_z \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u_z \quad \text{when firing the input } u_z. \tag{7}$$

With this equation in mind, we design here a hybrid feedback controller not only deciding the selection of the input u_z but also the triggering condition for firing the input, thereby inducing a jump as in (7). The proposed hybrid feedback controller comprises two internal states, one playing the role of a logic variable $q_z \in \{-1,1\}$ and one of them being a timer variable $\tau_z \in [0,2]$ whose value is constrained to evolve in a forward invariant compact set, selected as [0,2] by way of suitable scaling.

By gathering the overall out-of-plane closed-loop state in a vector $\xi_z := \begin{bmatrix} r_z & v_z & q_z & \tau_z \end{bmatrix}^\top$, and selecting $u_z = -v_z$, so that suitable damping of the closed-loop velocity is obtained in (7), we may write the closed-loop hybrid dynamics as

$$\begin{cases}
\begin{bmatrix} \dot{r}_z \\ \dot{v}_z \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -n^2 & 0 \end{bmatrix} \begin{bmatrix} r_z \\ v_z \end{bmatrix}, \\
\dot{\tau}_z = \frac{n}{2\pi} (1 - dz(\tau_z)), \\
\dot{q}_z = 0,
\end{cases} \qquad \xi_z \in \mathcal{C}_z, \tag{8a}$$

$$\begin{cases}
\begin{bmatrix} \dot{r}_z \\ \dot{v}_z \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -n^2 & 0 \end{bmatrix} \begin{bmatrix} r_z \\ v_z \end{bmatrix}, \\
\dot{\tau}_z = \frac{n}{2\pi} (1 - \operatorname{dz}(\tau_z)), \\
\dot{q}_z = 0,
\end{cases} \qquad \xi_z \in \mathcal{C}_z,$$

$$\begin{cases}
\begin{bmatrix} r_z^+ \\ v_z^+ \end{bmatrix} = \begin{bmatrix} r_z \\ v_z \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \operatorname{sat}(v_z), \\
\tau_z^+ = 0, \\
q_z^+ = -q,
\end{cases} \qquad \xi_z \in \mathcal{D}_z,$$
(8b)

where one clearly sees that the timer τ_z is reset to zero at each firing of the input $u_z = -\operatorname{sat}(v_z)$ and that it keeps track of each revolution of the oscillatory free (continuous) dynamics (6). In particular, after one revolution, unless it is reset before, the timer reaches the value $\tau_z = 1$. The dead-zone function, defined by $dz(u) = u - sat_1(u)$ with sat_1 being the symmetric saturation function with maximal amplitude 1, appearing in the flow dynamics of τ_z ensures that after $\tau_z \geq 1$, its speed is suitably slowed down until it is completely stopped once $\tau_z=2$. This ensures that the set [0,2] is forward invariant for τ_z , even though a consequence of this is that once $\tau_z \ge 1$ it does not anymore keep track of the revolutions of the continuous dynamics. About the logic variable q_z , note that it toggles between -1 and 1 at jumps. We may now define the flow and jump sets as given by

$$C_z = \left\{ \xi_z \in \mathbb{R}^3, r_z(v_z - nr_z) \le 0 \text{ or } q_z v_z \le 0 \text{ or } \tau_z \le \tau_z^M \right\},$$

$$\mathcal{D}_z = \left\{ \xi_z \in \mathbb{R}^3, r_z(v_z - nr_z) \ge 0 \text{ and } q_z v_z \ge 0 \text{ and } \tau_z \ge \tau_z^M \right\},$$
(9)

where τ_z^M is in (0,2]. The objective is to guarantee global asymptotic stability of the following set

$$\mathcal{A}_z := \{ \xi_z : \ r_z = v_z = 0, q_z \in \{-1, 1\}, \tau_z \in [0, 2] \}$$

$$\tag{10}$$

for the hybrid closed loop, as clarified din the next statement.

Theorem 1: The set A_z in (10) is globally asymptotically stable for the hybrid dynamics (8),(9).

Proof: The proof is based on the application of the La Salle invariance principle in [18, Thm 1] with the Lyapunov function given by $V(\xi_z) = n^2 r_z^2 + v_z^2$.

IV. STABILIZATION IN THE
$$(x, y)$$
 PLANE

In this section, the objective is to follow a similar path to the one laid down in Section III for the z direction, to solve the more challenging problem of stabilizing the coupled dynamics in the (x, y) plane. Combining these two controllers will lead to an overall hybrid stabilizing control system.

To this end, we introduce the following dynamics, describing the hybrid evolution in the (x, y) plane, as a function of the impulsive inputs u_x and u_y ,

$$\begin{bmatrix} \dot{r}_{x} \\ \dot{v}_{x} \\ \dot{r}_{y} \\ \dot{v}_{y} \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 1 & 0 & 0 \\ 3n^{2} & 0 & 0 & 2n \\ 0 & 0 & 0 & 1 \\ 0 & -2n & 0 & 0 \end{bmatrix}}_{=A_{0}} \begin{bmatrix} r_{x} \\ v_{x} \\ r_{y} \\ v_{y}^{+} \end{bmatrix}, \text{ in free motion}$$

$$= A_{0}$$

$$\begin{bmatrix} r_{x}^{+} \\ v_{x}^{+} \\ r_{y}^{+} \\ v_{y}^{+} \end{bmatrix} = \begin{bmatrix} r_{x} \\ v_{x} \\ r_{y}^{+} \\ v_{y}^{+} \end{bmatrix} + \underbrace{\begin{bmatrix} 0 & 0 \\ 1 & 0 \\ 0 & 1 \end{bmatrix}}_{=B_{0}} \begin{bmatrix} \text{sat}(u_{x}) \\ \text{sat}(u_{y}) \end{bmatrix}, \text{ when firing inputs}$$

$$= B_{0}$$

$$(11)$$

For obtaining an insightful representation of dynamics (11), introduce the coordinate transformation

$$\zeta = \begin{bmatrix} x \\ y \\ \alpha \\ \beta \end{bmatrix} := \underbrace{\begin{bmatrix} -3 & 0 & 0 - 2/n \\ 0 & 1 & 0 & 0 \\ 0 & -2/n & 1 & 0 \\ -6n & 0 & 0 & -3 \end{bmatrix}}_{:=T} \begin{bmatrix} r_x \\ v_x \\ r_y \\ v_y \end{bmatrix}$$
(12)

With these new coordinates, the dynamics (11) writes

$$\dot{\zeta} = A\zeta, \quad \text{in free motion}$$

$$\zeta^{+} = \zeta + B \begin{bmatrix} \operatorname{sat}(u_{x}) \\ \operatorname{sat}(u_{y}) \end{bmatrix}, \quad \text{when firing inputs}$$
(13)

with $A = TA_0T^{-1}$ and $B = TB_0$, corresponding to

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -n^2 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix}, \ B = \begin{bmatrix} 0 & -2/n \\ 1 & 0 \\ -2/n & 0 \\ 0 & -3 \end{bmatrix}$$

The structure of matrices A and B highlights that, along flowing solutions, the dynamics of the system is driven by two independent subsystems, one being an oscillator (having states x, y) and a second one being a double integrator (having states α, β). Across jumps, the structure of B introduces a coupling between both dynamics. Indeed, each impulsive action through u_x and u_y affects both the oscillator and the double integrator. Indeed, u_x affects y and α , while u_y affects y and β . To simplify the controller design, we exploit the structure of the system to provide a hierarchical control law to split the action of the two control inputs. More specifically, first the control input u_y is used to ensure finite-time convergence to zero of β . Then, the control input u_x is used to stabilize the remaining states $[x \ y \ \alpha]^{\top}$, following the same type of control laws as for the dynamics in the z direction, discussed in Section III. These two stabilizers are discussed in the next sections.

A. Finite-time stabilization of β

The proposed hybrid feedback controller for the dynamics of β comprises one internal state, being a timer variable $\tau_{\beta} \in [0,2]$ whose value is constrained to evolve in a forward invariant compact set, selected as [0,2] by way of a suitable scaling, for simplified notation. By gathering the plant-controller state in a vector $\xi_{\beta} := \begin{bmatrix} \beta & \tau_{\beta} \end{bmatrix}^{\top}$, and selecting $u_y = -\beta/3$, we may write the closed loop hybrid dynamics as follows

$$\begin{cases} \dot{\beta} = 0; \\ \dot{\tau}_{\beta} = \frac{n}{2\pi} (1 - dz(\tau_{\beta})) \end{cases} \xi_{\beta} \in \mathcal{C}_{\beta}$$
 (14a)

$$\begin{cases} \dot{\beta} = 0; \\ \dot{\tau}_{\beta} = \frac{n}{2\pi} (1 - dz(\tau_{\beta})) \end{cases} \qquad \xi_{\beta} \in \mathcal{C}_{\beta}$$

$$\begin{cases} \beta^{+} = \beta - 3 \operatorname{sat}(\beta/3) \\ \tau_{\beta}^{+} = 0, \end{cases} \qquad \xi_{\beta} \in \mathcal{D}_{\beta}.$$
(14a)

Mimicking the solution in Section III, this hybrid dynamics includes a timer τ_{β} , which is reset to zero at each firing of the input $u_y = \text{sat}(3\beta)$ and obeys flow dynamics ensuring that τ_β never leaves the compact set [0,2]. For a given $\tau_\beta^{\bar{M}}$ in (0,2], we may then define the flow and jump sets as follows

$$C_{\beta} = \left\{ \xi_{\beta} \in \mathbb{R}^2, \ \tau_{\beta} \le \tau_{\beta}^M \right\}, \ \mathcal{D}_{\beta} = \left\{ \xi_{\beta} \in \mathbb{R}^2, \ \tau_{\beta} \ge \tau_{\beta}^M \right\}, \tag{15}$$

The objective is now to guarantee global finite-time stability of the following set

$$A_{\beta} := \{ \xi_{\beta} : \beta = 0, \, \tau_{\beta} \in [0, 2] \}$$
 (16)

for the hybrid closed loop, as clarified in the next statement. We recall that finite-time stability comprises global asymptotic stability and finite-time convergence.

Theorem 2: The set A_{β} in (16) is finite-time stable for the hybrid dynamics (14),(15).

Proof: The proof is based on the Lyapunov function given by $V(\xi_{\beta}) = \beta^2$.

Remark 2: Note that the definition of the flow and jump sets \mathcal{C}_{β} and \mathcal{D}_{β} imposes periodic impulses. Such a periodic behavior could have been modelled using a simple timer instead of this more involved solution. This choice is made to be consistent with the other timers required for the two other hybrid controller.

Remark 3: Even though the control input u_y is triggered periodically, the finite convergence of β to the origin ensures that the magnitude of the control law $sat(\beta/3)$ will also be zero in finite time, and no control action will be applied.

B. Stabilization of x, y, α

Following the previous paragraph, a first layer of the control law suitably selects the control input u_y to ensure finite-time convergence to zero of the variable β , in addition to finite-time convergence to zero of u_u itself. In this section, we will exploit a suitable hybrid feedback selection of u_x to stabilize the variables x, y and α to the origin. Once β and u_y have converged to zero, we may design a feedback controller based on a logic variable $q_{\alpha} \in \{-1,1\}$ and on a time variable $\tau_{\alpha} \in [0,2]$, by focusing on the following reduced dynamics with plant states $[x \ y \ \alpha]^{\top}$ and controllers states q_{α} and τ_{α} gathered in an overall vector ξ_{α} ,

$$\begin{cases}
\begin{bmatrix} \dot{x} \\ \dot{y} \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ -n^2 & 0 & 0 \end{bmatrix} \begin{bmatrix} x \\ y \\ \alpha \end{bmatrix} \\
\dot{q}_{\alpha} = 0 \\
\dot{\tau}_{\alpha} = \frac{n}{2\pi} (1 - dz(\tau_{\alpha}))
\end{cases} \qquad \xi_{\alpha} \in \mathcal{C}_{\alpha}, \tag{17a}$$

$$\begin{cases}
\begin{bmatrix} \dot{x} \\ \dot{y} \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ -n^2 & 0 & 0 \end{bmatrix} \begin{bmatrix} x \\ y \\ \alpha \end{bmatrix} \\
\dot{q}_{\alpha} = 0 \\
\dot{\tau}_{\alpha} = \frac{n}{2\pi} (1 - dz(\tau_{\alpha}))
\end{cases}$$

$$\begin{cases}
\begin{bmatrix} x^+ \\ y^+ \\ \alpha^+ \end{bmatrix} = \begin{bmatrix} x \\ y \\ \alpha \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \\ -\frac{2}{n} \end{bmatrix} sat(u_x) \\
q^+_{\alpha} = -q_{\alpha} \\
\tau^+_{\alpha} = 0,
\end{cases}$$

$$(17a)$$

Following the control design of the dynamics in the z direction (see Section III, we use the following hybrid control law for (17), consisting in the control input $u_x = \frac{n\alpha}{4} - \frac{y}{2}$, with

$$\mathcal{C}_{\alpha} = \begin{cases}
\xi_{\alpha} \in \mathbb{R}^{3}, & \text{or} & (y - \frac{n}{2}\alpha - nx)x \leq 0, \\
q_{\alpha}(y - \frac{n}{2}\alpha) \leq 0, \\
\tau_{\alpha} \leq \tau_{\alpha}^{M}
\end{cases},$$

$$\tau_{\alpha} \leq \tau_{\alpha}^{M}$$

$$\mathcal{D}_{\alpha} = \begin{cases}
\xi_{\alpha} \in \mathbb{R}^{3}, & \text{and} & (y - \frac{n}{2}\alpha - nx)x \geq 0, \\
q_{\alpha}(y - \frac{n}{2}\alpha) \geq 0, \\
\text{and} & \tau_{\alpha} \geq \tau_{\alpha}^{M}.
\end{cases}$$
(18)

for a given au_{α}^M in (0,2]. This feedback system ensures global asymptotic stability of the following set

$$\mathcal{A}_{\alpha} := \left\{ \xi_{\alpha} \in \mathbb{R}^3 : \ \xi_{\alpha} = 0, \ q_{\alpha} \in \{-1, 1\}, \ \tau_{\alpha} \in [0, 2] \right\}, \tag{19}$$

as clarified din the next statement.

Theorem 3: The set A_{α} in (19) is globally asymptotically stable for the hybrid dynamics (17),(18).

Proof: The proof is based on the application of the La Salle invariance principle [12, Chapter 8] with the quadratic Lyapunov function $V_{\alpha}(\xi_{\alpha}) = n^2 x^2 + y^2 + \frac{n^2}{4} \alpha^2$.

V. COMBINED HYBRID FEEDBACK

In Sections III and IV we have designed three nested hybrid controllers, each of them inducing desirable properties of the corresponding dynamics. Their hybrid combination is discussed here.

To combine the three controllers, let us first define the overall state $\xi = [\xi_z^\top \ \xi_\beta^\top \ \xi_\alpha^\top]^\top$ for the closed-loop dynamics:

$$\xi = [r_z \ v_z \ \tau_z \ q_z \ \beta \ \tau_\beta \ x \ y \ \alpha \ q_\alpha \ \tau_\alpha]^\top. \tag{20}$$

State ξ clearly evolves in the following set:

$$\mathbb{X} = \mathbb{R}^2 \times [0, 2] \times \mathcal{Q} \times \mathbb{R} \times [0, 2] \times \mathbb{R}^3 \times \mathcal{Q} \times [0, 2], \tag{21}$$

where we denoted $Q = \{-1, 1\}$. Based on state ξ , we may define the following extended selections of the impulsive feedback control law:

$$\kappa_z(\xi) = \begin{bmatrix} 0 \\ u_z \end{bmatrix}; \ \kappa_\beta(\xi) = \begin{bmatrix} 0 \\ u_y \\ 0 \end{bmatrix}; \ \kappa_\alpha(\xi) = \begin{bmatrix} u_x \\ 0 \\ 0 \end{bmatrix}, \tag{22}$$

with $u_z=-v_z,\,u_y=-\beta/3$ and $u_x=\frac{n\alpha}{4}-\frac{y}{2}$ as per the selections in (8), (14), and (17). We may then just as well define the three following jump maps and jump sets, each of them characterizing the corresponding stabilizer, whose properties have been established in Theorems 1, 2 and 3:

$$g_{z}(\xi) = \begin{bmatrix} r \\ v + \operatorname{sat}(\kappa_{z}(\xi)) \end{bmatrix}, \quad \xi \in \overline{\mathcal{D}}_{z} := \{ \xi : \xi_{z} \in \mathcal{D}_{z} \}$$

$$g_{\beta}(\xi) = \begin{bmatrix} r \\ v + \operatorname{sat}(\kappa_{\beta}(\xi)) \end{bmatrix}, \quad \xi \in \overline{\mathcal{D}}_{\beta} := \{ \xi : \xi_{\beta} \in \mathcal{D}_{\beta} \}$$

$$g_{\alpha}(\xi) = \begin{bmatrix} r \\ v + \operatorname{sat}(\kappa_{\alpha}(\xi)) \end{bmatrix}, \quad \xi \in \overline{\mathcal{D}}_{\alpha} := \{ \xi : \xi_{\alpha} \in \mathcal{D}_{\alpha} \}.$$

The overall control scheme is then the one prioritizing jumps, accounting for the three selections above. We may select the jump set $\mathcal{D} \subset \mathbb{X}$ and the flow set $\mathcal{C} \subset \mathbb{X}$ as

$$\mathcal{D} = \overline{\mathcal{D}}_z \cup \overline{\mathcal{D}}_\beta \cup \overline{\mathcal{D}}_\alpha, \quad \mathcal{C} = \overline{\mathbb{X} \setminus \mathcal{D}}, \tag{23}$$

where the over-line denotes the closure operation, so that the flow set is the closed complement of the jump set.

As for the jump map, following standard practices, we select it as a set-valued jump map G whose graph is the union of the graphs of g_z , g_β and g_α defined above. More specifically, G contains all the possible update laws included in the three pairs $(\overline{\mathcal{D}}_z, g_z)$, $(\overline{\mathcal{D}}_\beta, g_\beta)$, and $(\overline{\mathcal{D}}_\alpha, g_\alpha)$ so that, as an example, when $\xi \in (\overline{\mathcal{D}}_z \cap \overline{\mathcal{D}}_\beta) \setminus \overline{\mathcal{D}}_\alpha$, then $G(\xi) = \{g_z(\xi)\} \cup \{g_\beta(\xi)\}$, and similarly for the other cases.

Denoting by f(x) the juxtaposition of the flow maps in (8), (14) and (17), respectively, the overall closed loop writes:

$$\dot{\xi} = f(\xi), \qquad \qquad \xi \in \mathcal{C} \tag{24a}$$

$$\xi^+ \in G(\xi),$$
 $\xi \in \mathcal{D}.$ (24b)

The following result states our result about stability properties of the following set for the closed-loop dynamics (24),

$$A = \{ \xi : r_z = v_z = 0, x = y = 0, \alpha = \beta = 0 \}.$$
(25)

Theorem 4: The set A is globally asymptotically stable for the overall closed-loop system (24).

Proof: The proof is based on a reduction argument from [19, Cor. 4.8] and is omitted here due to space limitation. ■

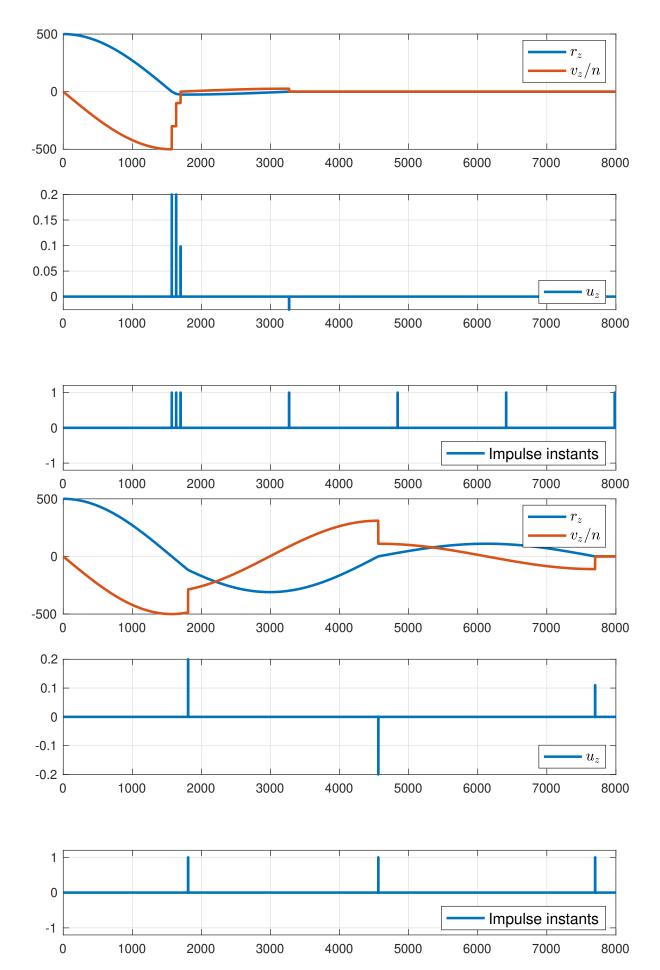


Fig. 2. Simulation of system (8) with $\tau_z^M=0.01$ (left) and $\tau_z^M=0.25$ (right). The figure shows the evolution of the state variables r_z,v_z (top), the magnitude of the control input (middle) and the instants of impulses (bottom).

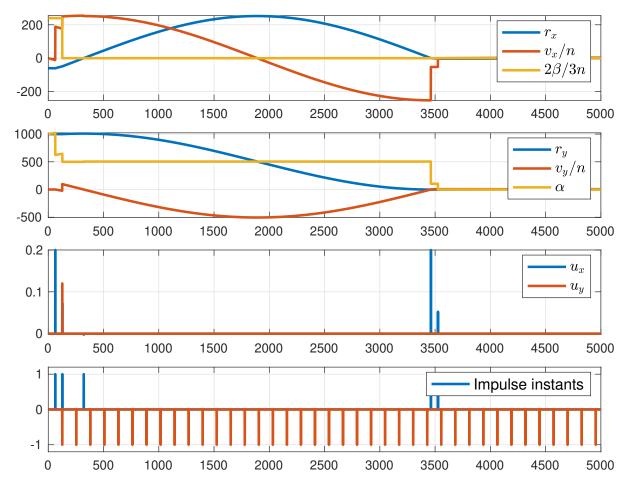


Fig. 3. Evolution in the original coordinates $r_x, v_x/n$ with the transformed state $2\beta/3$ from (12) (top plot), as well as the original coordinates $r_y, v_y/n$ together with the transformed state α (second plot). Magnitude of the control input (third plot) and instants of impulses (bottom plot).

VI. SIMULATION RESULTS

This section aims to illustrate the numerical application of the proposed saturated control laws with a symmetric saturation level chosen as $u_{\rm M}=0.2m/s$. Figure 2 presents two simulations for the closed-loop system (8) with two different values of the dwell time parameter τ_z^M . In both cases, the state variables (r_z,v_z) converge asymptotically to the origin. These figures also demonstrate that our control triggers impulses when r_z crosses zero, the most efficient situation for the control action. When $\tau_z<0.25$ (a quarter of a rotation) an additional impulse is triggered to compensate for the saturation. Both cases show a trade-off between the rate of convergence (faster with $\tau_z^M=0.01$) and the cost of consumption (lower with $\tau_z^M=0.25$, i.e., only three impulses).

 $au_z^M=0.25$, i.e., only three impulses). Figures 3 and 4 depict a simulation of the system (14), (17) (in-plane) with the dwell-time parameters $au_\alpha^M=0.01$ and $au_\beta^M=0.02$, along with the initial condition $[-60\ 0\ 1000\ 0]^{\rm T}$. This simulation provides an overview of the entire dynamics in the (x,y)-plane, which includes the closed-loop systems (14) (associated with state ξ_β) and (17) (associated with state ξ_α). The bottom plot in Figure 3 shows the periodic impulses (in red) generated by system (14). As mentioned earlier in the paper, although the system regularly enters the jump set, no control action is required after a while, since the variable β reaches zero after only one impulse (see the red impulse u_y in the third plot).

The two graphs at the top of Figure 3 demonstrate that the trajectory of the closed-loop systems (14), (17) converges asymptotically to the origin. Since the dwell time parameter τ_{α}^{M} is less than 0.25, the control law allows for a series of three successive impulses to compensate for the effects of saturation at the maximum amplitude $u_{\rm M}$ and achieve a fast convergence rate. As for the z axis, selecting larger values of τ_{α}^{M} (larger than 0.25), it is possible to reduce the number of impulses (consumption), but at the cost of deteriorating the convergence rate.

Finally, Figure 4 illustrates the trajectory of the closed-loop systems in the (r_x, r_y) plane (bottom). The seemingly simple behavior of the resulting solution hides a non-trivial hybrid trajectory of variables (x, y) depicted in the top graph, which experiences several jumps throughout the simulation.

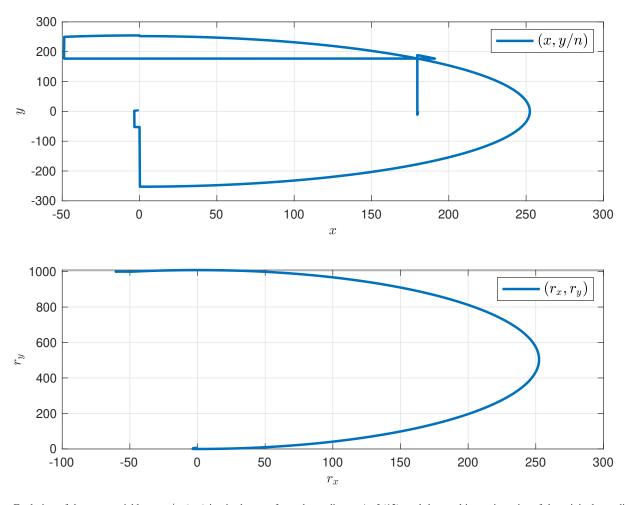


Fig. 4. Evolution of the state variables x, y/n (top) i.e. in the transformed coordinates ζ of (12), and the resulting trajectories of the original coordinates (r_x, r_y) (bottom).

VII. CONCLUSION

This paper presented a hybrid dynamical system approach for the impulsive control in spacecraft rendezvous and proximity operations, using the Hill-Clohessy-Wiltshire model. The control design problem was tackled by separating the out-of-plane from the in-plane dynamics, with a distinct feedback control law developed for each one of them. These laws were grounded on specially designed Lyapunov functions, accounting for thrusters saturation. Future work will include addressing minimum impulse bit requirements, in addition to safety constraints and the use of these control laws in more complex rendezvous scenarios, such as formation-flying (multiple spacecraft rendezvous), eccentric orbits (time-varying dynamics) and Halo orbit rendezvous (highly nonlinear dynamics). In addition the performance our hybrid control design should be compared and evaluated with other approaches (e.g. MPC) in a high-fidelity simulator.

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