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► **To cite this version:**

Zhijian Hu, Qingyang Li, Pu Zhang, Ruiping Wang, Kai Zhang. A Novel Handling Method to Intermittent Feedback in Load Frequency Regulation for Renewable Energy-Dominated Microgrids. IEEE Transactions on Instrumentation and Measurement, 2024, 73, 10.1109/TIM.2024.3457958 . hal-04732942

**HAL Id: hal-04732942**

**<https://laas.hal.science/hal-04732942v1>**

Submitted on 11 Oct 2024

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# A Novel Handling Method to Intermittent Feedback in Load Frequency Regulation for Renewable Energy-Dominated Microgrids

Zhijian Hu, *Member, IEEE*, Qingyang Li, Pu Zhang, *Member, IEEE*, Ruiping Wang, and Kai Zhang

**Abstract**—Traditional thermal-based microgrids are transforming towards distributed renewable energy sources (DRESSs) dominated microgrids, aiming at promoting sustainable environment and economy. This work thus incorporates two popular DRESSs (i.e., wind turbines and plug-in electric vehicles (PEVs)) in the thermal-based power plants. Considering the inevitable sensor failures caused by hostile outdoor environment and the communication failures caused by resource constraints, this paper proposes a novel handling method to intermittent feedback in load frequency control (LFC) for microgrids. Furthermore, to deal with the fluctuations of power outputs from DRESSs, this paper incorporates the robust performance index into the control law derivation. By taking into account such intermittent feedback and robust performance index, the operation of microgrid system can better realize resiliency enhancement. Validation results demonstrate the feasibility of this novel handling method under different degree of intermittent feedback.

**Index Terms**—Intermittent measurement, microgrid, load frequency control, wind turbine, plug-in electric vehicle.

## I. INTRODUCTION

Growing climate changes put forward higher requirements to energy sectors to promote sustainable development and decarbonization. Increasing the penetration level and the diversity of distributed renewable energy sources (DRESSs) into the traditional thermal-based microgrids becomes a promising solution [1], [2]. While, with the increasing penetration of DRESSs, the physical inertia of power systems, traditionally provided by synchronous generators, reduces sharply. This incurs greater challenges to frequency stability guarantees.

This work is supported in part by the European Union’s Horizon 2022 Research and Innovation Programme for the Marie Skłodowska-Curie Actions under Grant 101108472, in part by National Natural Science Foundation of China under Grant 62203142, in part by Aeronautical Science Foundation of China under Grant 2022Z071077009, and in part by New Era-Longjiang Excellent Master and Ph.D Thesis Funding Project under Grant LJYXL2022-008. (*Corresponding author: Kai Zhang.*)

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Furthermore, the renewable power outputs of DRESSs are fluctuant due to the uncertainties of natural resources, such as the stochastic wind speed. As a result, this paper aims to incorporate multiple DRESSs, and in this context, propose a robust frequency controller to reduce the influence of DRESSs power output fluctuation on frequency stability.

Among DRESSs, the wind power generation has emerged as a most critical contributor to promote the green energy exploration, as of the most recent assessments in 2023 [3]. The modular scalability of wind farms, coupled with advancements in turbine technology and economies of scale, has contributed to the increasing cost-effectiveness of wind energy. Research and development initiatives have furthered the understanding of wind dynamics [4], aiding in advancing wind turbine control technologies for stable wind power harvesting. Despite notable progress has been made to understand the wind turbine model and the wind power profile, the application challenges such as fluctuant power output and grid integration persist, necessitating the effectiveness explorations of existing control methodologies in real-world scenarios.

For load frequency control (LFC) design at the microgrid level, consecutive feedback control signals are a prerequisite to ensure the frequency stability [5], [6]. On the one hand, due to the fact that smart sensors (e.g., the phase measurement units (PMUs) and remote telemetry units (RTUs)) are usually allocated in outdoor environment [7], they potentially suffer functional faults such as failed element, degraded processing capacity, and sensor aging. As a direct consequence of sensor faults, measurements from sensors are intermittent, which disable the LFC implementation with output feedback control techniques. On the other hand, modern microgrids prefer the wireless sensor network (WSN) to realize data sharing, during which process the communication link failure may occur [8]–[10]. However, existing results addressing multiple sources of intermittent feedback control signals within sensor-controller loop for LFC issues are limited. Considering the intermittent measurements from sensors, the authors in [11]–[19] revealed some effective solutions. From the perspective of modeling, the authors in [11] took into account PMU failures caused by disorder interaction between hardware and software and modeled this category of failures using a Markov model-based unified PMU reliability model, which greatly aided in the estimation of PMU failure data using the Monte Carlo simulation technique. The authors in [15] characterized PMU failure as a hidden Markov model and performed the reliability analysis of PMUs, based on which the transient probability is computed,



dynamics is described as

$$\dot{i}_{qr} = -\left(\frac{1}{T_1}\right)i_{qr} + \left(\frac{X_2}{T_1}\right)V_{qr}, \quad (1)$$

$$\dot{\omega} = -\left(\frac{X_3}{2M_t}\right)i_{qr} + \left(\frac{1}{2M_t}\right)T_m, \quad (2)$$

where  $X_2 = 1/R_r$ ,  $X_3 = L_m/L_{ss}$ ,  $T_1 = L_0/(\omega_s R_s)$ ,  $T_e = X_3 i_{qr}$ ,  $L_0 = L_{rr} + L_m^2/L_{ss}$ ,  $L_{ss} = L_s + L_m$ ,  $L_{rr} = L_r + L_m$ ;  $\omega$  is the rotational speed;  $T_m$  is the mechanical torque;  $P_W = \omega T_e$  is the active power output;  $i_{qr}$  and  $V_{qr}$  are the  $q$ -axial rotor current and voltage; meanings of other parameters in (1) and (2) can be found in [26].

By choosing  $x_W = [i_{qr} \ \omega]^T$ , (1) and (2) can be formulated as

$$\begin{cases} \dot{x}_W = \mathcal{A}_W x_W + \mathcal{B}_W u_W + \mathcal{F}_W v_W, \\ y_W = \mathcal{C}_W x_W, \end{cases} \quad (3)$$

where  $u_W = V_{qr}$  is the control action;  $v_W = T_m$  is the electromagnetic disturbance;  $y_W$  is the measurement with  $\mathcal{C}_W = I_2$ ;

$$\mathcal{A}_W = \begin{bmatrix} \frac{-1}{T_1} & 0 \\ -X_3 & 0 \end{bmatrix}, \mathcal{B}_W = \begin{bmatrix} \frac{X_2}{T_1} \\ 0 \end{bmatrix}, \mathcal{F}_W = \begin{bmatrix} 0 \\ 1 \\ \frac{1}{2M_t} \end{bmatrix}.$$

In view that the state measurements of wind power generation systems are usually realized by digital sensors to facilitate the control design, (3) is discretized to

$$x_W(k+1) = A_W x_W(k) + B_W u_W(k) + F_W v_W(k), \quad (4)$$

$$y_W(k) = C_W x_W(k), \quad (5)$$

where  $A_W = e^{\mathcal{A}_W l}$ ,  $B_W = \int_0^l e^{\mathcal{A}_W l} \mathcal{B}_W dl$ ,  $C_W = \mathcal{C}_W$ , and  $F_W = \int_0^l e^{\mathcal{A}_W l} \mathcal{F}_W dl$ ;  $l$  is the sampling period.

### B. MPC Formulation

To actualize the assigned power tracking from wind turbines, this work employs the MPC policy, a most promising approach in microgrids due to its superiority in solving finite-horizon constrained optimization problems at each sampling instant. The cost function of the MPC is chosen as

$$\begin{aligned} \mathcal{J}(\mathbf{x}_W(k), \mathbf{u}_W(k)) &= \sum_{m=0}^{\bar{M}-1} (\|x_W(k+m|k) - x_W^{\text{ref}}\|_Q^2 \\ &+ \|u_W(k+m|k) - u_W^{\text{ref}}\|_S^2) + V_f(x_W(k+\bar{M}|k)) \end{aligned} \quad (6)$$

where  $\mathbf{u}_W(k) = [u_W(k|k)^T \ u_W(k+1|k)^T \ \dots \ u_W(k+\bar{M}-1|k)^T]^T$ , and  $\mathbf{x}_W(k) = [x_W(k+1|k)^T \ x_W(k+2|k)^T \ \dots \ x_W(k+\bar{M}|k)^T]^T$ ;  $V_f = x_W(k+\bar{M}|k)^T P x_W(k+\bar{M}|k)$  is the terminal cost related to system stability;  $\bar{M}$  defines the prediction (control) horizon;  $x_W^{\text{ref}}$  is the reference of  $x_W$ ;  $u_W^{\text{ref}}$  is the reference of  $u_W$ ;  $k+m|k$  means the predicted value at time  $k+m$  with the real-time measurements at time  $k$ ;  $Q \succ 0$  and  $S \succeq 0$  are two tuning matrices. The selection of control and prediction horizons necessitates a careful balance between tracking accuracy and computational constraints. To enhance the tracking precision and effective disturbance rejection, a

longer prediction horizon is advisable. However, this increases computational complexity, potentially impeding real-time implementation. Therefore, it is imperative to ensure that the selected prediction horizon does not introduce undue delays. Besides, the control horizon should be appropriately set to be shorter than the prediction horizon, optimizing computational efficiency while maintaining control performance.

The MPC approach can well address states constraints during the process of seeking for optimal control actions. For wind power generation systems, the constraints mainly refer to current, rotational speed, stator/rotor side converters. Considering these constraints, a formal MPC policy for assigned power tracking is

$$\begin{aligned} &\min_{\mathbf{u}_W} \mathcal{J}(\mathbf{x}_W(k), \mathbf{u}_W(k)) \\ &\text{s.t.} \begin{cases} x_W(k+m+1|k) = A_W x_W(k+m|k) \\ \quad + B_W u_W(k+m|k), \\ x_W(k|k) = x_W(k), \\ x_W(k+\bar{M}|k) \in \Omega, \\ i_{qr,\min} \leq i_{qr}(k+m|k) \leq i_{qr,\max}, \\ \omega_{\min} \leq \omega(k+m|k) \leq \omega_{\max}, \\ m = 0, 1, 2, \dots, \bar{M}-1, \end{cases} \end{aligned} \quad (7)$$

where  $\Omega$  is the terminal region constraint;  $i_{qr,\min}$  ( $i_{qr,\max}$ ) and  $\omega_{\min}$  ( $\omega_{\max}$ ) are the lower (upper) bounds of  $i_{qr}$  and  $\omega$ .

## III. RESILIENT LFC CONSIDERING INTERMITTENT FEEDBACK SIGNALS

### A. Modeling of PEVs

The power source of PEVs is the battery, which is typically constituted by a multitude of interconnected battery cells. They establish a connection with the power grid through a DC/AC converter [27]. For integration with the power grid, the equivalent circuit is shown in Fig. 2(a), where  $I_{dc}$  and  $V_{dc}$  represent the battery's DC current and voltage, and  $P_{V2G}$  signifies the active power injected into the power grid via the vehicle-to-grid infrastructure.

Corresponding to Fig. 2(b), the system dynamics of the battery are represented by

$$-\frac{V_{boc} + V_{dc} + V_b}{R_{bc}} = I_{dc}, \quad (8)$$

$$C_{bp} \frac{dV_{boc}}{dt} + \frac{V_{boc}}{R_{bp}} = I_{dc}, \quad (9)$$

$$C_b \frac{dV_b}{dt} + \frac{V_b}{R_b} = I_{dc}, \quad (10)$$

where  $V_{boc}$  denotes the open circuit voltage;  $V_b$  signifies the overvoltage;  $R_{bp}$  represents the resistance of self discharging;  $R_b$  and  $C_b$  account for the transient overvoltage effects;  $C_{bp}$  means the capacitance; and  $R_{bc}$  stands for the resistance related to the terminals and inter-cell connections of the battery.

By ignoring the power loss of the equivalent circuit, the power supplied to the grid is identical to that generated by the battery, which can be represented by

$$P_{V2G} = P_{dc} = I_{dc} V_{dc}. \quad (11)$$

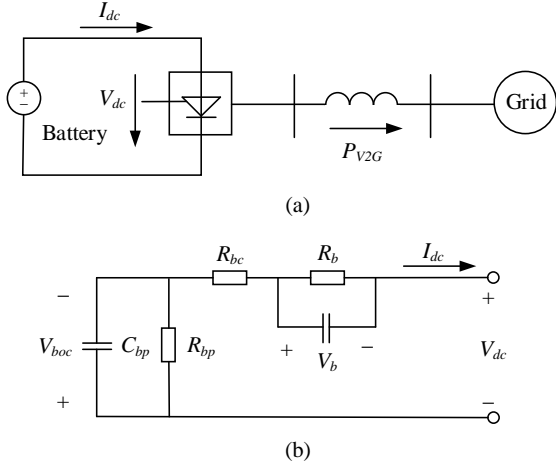


Fig. 2. PEV model [27]. (a) PEV's equivalent circuit; (b) Battery's equivalent circuit.

By neglecting the higher-order components, the increment of V2G power from (11) is calculated by

$$\Delta P_{V2G} = V_{dc}^0 \Delta I_{dc} + I_{dc}^0 \Delta V_{dc}, \quad (12)$$

where  $\Delta V_{dc} = V_{dc} - V_{dc}^0$ ,  $\Delta I_{dc} = I_{dc} - I_{dc}^0$ ; and  $V_{dc}^0$  and  $I_{dc}^0$  signify the original values of  $V_{dc}$  and  $I_{dc}$ .

Through dividing the DC voltage into two elements as

$$\Delta V_{dc} = \Delta V_r + \Delta V_s, \quad (13)$$

we can ultimately arrive at

$$\Delta P_{V2G} = V_{dc}^0 \Delta I_{dc} + I_{dc}^0 (\Delta V_r + \Delta V_s). \quad (14)$$

In light of the compensation aspect as explicated by the authors in [28], an incremental adjustment of the battery voltage is incorporated as a corrective measure to mitigate power fluctuations arising from deviations in current. Consequently, the term  $I_{dc}^0 \Delta V_r$  in (14) is specifically allocated for power compensation to alterations in  $\Delta I_{dc}$ . Concurrently, the term  $I_{dc}^0 \Delta V_s$  in (14) is attributed to the accommodation of V2G power in response to LFC. Subsequently, we have

$$\Delta P_{V2G} = I_{dc}^0 \Delta V_s. \quad (15)$$

Within LFC,  $\Delta V_s$  is meticulously modulated in response to variations in frequency, thereby yielding

$$\Delta \dot{V}_s = \frac{1}{T_b} (-\Delta V_s - k_v \Delta f). \quad (16)$$

As a result, the V2G power that one PEV contributes to LFC is

$$\Delta \dot{P}_{V2G} = \frac{1}{T_b} (-\Delta P_{V2G} - K_b \Delta f). \quad (17)$$

where  $K_b$  and  $T_b$  indicate the gain and the time constant of a battery; and  $K_b = k_v I_{dc}^0$ .

In real-world scenarios, an individual PEV can hardly participate in LFC due to the limited power capacity of the battery. Therefore, an aggregator, functioning as an intermediate agent between the PEVs and the grid, is usually employed

to aggregate the grid-scale power capacity for LFC from thousands of PEVs. As for the aggregation algorithm design, interested readers can refer to our previous work [29], where we proposed a maximal power acquisition algorithm tailored for PEV remaining capacity aggregation. To focus on the resilient LFC design under intermittent feedback signals, this work concisely substitutes  $\Delta P_{V2G}$  by  $\Delta P_{agg}$  in the following discussion.

## B. Modeling of Microgrids

The microgrid employed in this work (see Fig. 1) is made up of generator dynamic, governor dynamic, and microgrid dynamic. The system dynamics are described as

$$\Delta \dot{P}_m = -\frac{1}{T_d} \Delta P_m + \frac{K_d}{T_d} \Delta P_v, \quad (18)$$

$$\Delta \dot{P}_v = -\frac{K_g}{RT_g} \Delta f - \frac{1}{T_g} \Delta P_v + \frac{K_g}{T_g} \Delta P_c, \quad (19)$$

$$\begin{aligned} \Delta \dot{f} = & -\frac{1}{T_p} \Delta f + \frac{K_p}{T_p} \Delta P_m + \frac{K_p}{T_p} \Delta P_W - \frac{K_p}{T_p} \Delta P_L \\ & + \frac{K_p}{T_p} \Delta P_O + \frac{K_p}{T_p} \Delta P_{agg}, \end{aligned} \quad (20)$$

where  $K_g$ ,  $K_d$ , and  $K_p$  denote the governor, generator, and microgrid gains;  $\Delta P_W$  denotes the wind power deviation;  $\Delta P_O$  indicates the total deviations caused by other DRESs; other parameters are defined in [18].

Notably, (17)-(20) constitute the small-signal model for LFC by linearizing the power generation system around one steady-state operating point, wherein system parameters are only accurate for small load disturbances. For the multi-energy systems, the linearized system parameters may not always precise because of the potential movement of the steady-state operating point because of fluctuant power outputs from DRESs. To advance [22] where only the uncertainty on  $R$  is addressed, the paper addresses more general parameter uncertainties and describes them by incorporating an unknown and bounded matrix  $\Delta \mathcal{A}(t)$  to the existing small-signal model, which is capable to address all parameter uncertainties. Thus, the improved small-signal model for (17)-(20) becomes

$$\begin{cases} \dot{x} = (\mathcal{A} + \Delta \mathcal{A}(t))x + \mathcal{B}u + \mathcal{F}v, \\ y = \mathcal{C}x, \end{cases} \quad (21)$$

where  $x = [\Delta f \quad \Delta P_m \quad \Delta P_v \quad \Delta P_{agg}]^T$ ,  $\mathcal{C} = I_3$ ,  $u = \Delta P_c$ ;  $v = [\Delta P_W \quad \Delta P_O \quad \Delta P_L]^T$ ,  $v \in l_2[0, \infty)$  is generally seen as the load disturbance;  $\Delta \mathcal{A}(t)$  is an unknown and time-

varying matrix to explain the parameter change; and

$$\mathcal{A} = \begin{bmatrix} \frac{-1}{T_p} & \frac{K_p}{T_p} & 0 & \frac{1}{T_p} \\ 0 & \frac{-1}{T_d} & \frac{K_d}{T_d} & 0 \\ -K_g & 0 & \frac{-1}{T_g} & 0 \\ \frac{RT_g}{T_b} & 0 & 0 & \frac{-1}{T_b} \\ -K_b & 0 & 0 & \frac{-1}{T_b} \end{bmatrix},$$

$$\mathcal{F} = \begin{bmatrix} \frac{K_p}{T_p} & -\frac{K_p}{T_p} & \frac{K_p}{T_p} \\ 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}, \mathcal{B} = \begin{bmatrix} 0 \\ 0 \\ \frac{K_g}{T_g} \\ 0 \end{bmatrix}.$$

In LFC, all states are measured and transmitted by digital sensors, such as PMUs and RTUs. Hence, (21) is discretized by

$$\begin{cases} x(k+1) = (A + \Delta A(k))x(k) + Bu(k) + Fv(k), \\ y(k) = Cx(k), \end{cases} \quad (22)$$

where  $A = e^{\mathcal{A}l}$ ,  $B = \int_0^l e^{\mathcal{A}\tau} \mathcal{B} d\tau$ ,  $C = \mathcal{C}$ ,  $F = \int_0^l e^{\mathcal{A}\tau} \mathcal{F} d\tau$ , and  $\Delta A(k) = NG(k)Y$ ;  $l$  is the sampling time;  $N$  and  $Y$  are given matrices;  $G(k)$  is assumed to be an unknown, time-varying, and Lebesgue measurable matrix satisfying  $G^T(k)G(k) \leq I$ .

### C. Control Formulation with Intermittent Feedback Signals

A typical output feedback law for LFC is adopted,

$$u(k) = K\tilde{y}(k), \quad (23)$$

where  $K$  is the controller gain to be determined, and  $\tilde{y}(k)$  is the actual signal used in control formulation.

To formulate a more practical frequency controller, this work considers the intermittent feedback signals, include the intermittent measurement caused by temporary sensor failures and the intermittent transmission caused by unreliable communication networks. To model the temporary sensor failures, this work employs Bernoulli variable  $\vartheta(k)$  [17], [18], which is able of capturing the intermittency extent by flexibly selecting various probability distributions. The actual measurement  $\tilde{y}(k)$  from the sensor is depicted by

$$\tilde{y}(k) = \vartheta(k)y(k), \quad (24)$$

where  $\text{Prob}\{\vartheta(k) = 1\} = \bar{\vartheta}$ ,  $\text{Prob}\{\vartheta(k) = 0\} = 1 - \bar{\vartheta}$  with  $\bar{\vartheta} \in [0, 1]$  indicating the probability that the signal  $y(k)$  is accurately measured, and  $\text{Prob}\{\cdot\}$  indicating the probability operator.

To model the intermittent transmission caused by unreliable communication networks, the Bernoulli variable  $\varrho(k)$  is adopted. The practically transmitted signal to the control center is depicted as

$$\tilde{y}(k) = \varrho(k)\tilde{y}(k), \quad (25)$$

where  $\text{Prob}\{\varrho(k) = 1\} = \bar{\varrho}$ ,  $\text{Prob}\{\varrho(k) = 0\} = 1 - \bar{\varrho}$  with  $\bar{\varrho} \in [0, 1]$  representing the probability that the measurement

is successfully transmitted from sensor side to the controller side.

By taking (23)-(25) to (22), the closed-loop LFC model is deduced as,

$$x(k+1) = (A + \Delta A(k) + \vartheta(k)\varrho(k)BKC)x(k) + Fv(k). \quad (26)$$

### D. Stability Analysis

To promote the control law derivation, we propose Theorem 1. Based on the sufficient condition therein, we can automatically calculate the resilient controller gains by resorting to the mincx solver in linear matrix inequality toolbox (LMI) in Matlab.

*Theorem 1:* Considering the intermittent measurement probability  $\bar{\vartheta}$  and the intermittent transmission probability  $\bar{\varrho}$ , the system (26) is mean-square asymptotically stable with  $\Delta f$  satisfying the predefined  $\mathcal{H}_\infty$  index  $\gamma$  if there exists matrix  $P \succ 0$ , such that

$$\begin{bmatrix} \tilde{\Omega} + D^T D & \tilde{A}^T P F \\ \star & F^T P F - \gamma^2 I \end{bmatrix} \prec 0, \quad (27)$$

where  $\tilde{\Omega} = \tilde{A}^T P \tilde{A} + \tilde{B}^T P \tilde{B} - P$ ,  $\tilde{B} = \sqrt{\bar{\vartheta}\bar{\varrho}(1 - \bar{\vartheta}\bar{\varrho})}BKC$ ,  $\tilde{A} = A + \Delta A(k) + \bar{\vartheta}\bar{\varrho}BKC$ , and  $D = \text{diag}\{1, 0, 0, 0\}$  with  $\text{diag}\{\cdot\}$ .

*Proof:* By choosing the following Lyapunov function

$$V(k) = x^T(k)Px(k),$$

one can derive

$$\begin{aligned} \Delta V(k) &= \mathcal{E}\{[x^T(k)(\tilde{A} + (\vartheta(k)\varrho(k) - \bar{\vartheta}\bar{\varrho})BKC)^T \\ &\quad + v^T(k)F^T]P[(\tilde{A} + (\vartheta(k)\varrho(k) - \bar{\vartheta}\bar{\varrho})BKC)x(k) \\ &\quad + Fv(k)]\} - x^T(k)Px(k) \\ &= \begin{bmatrix} x(k) \\ v(k) \end{bmatrix}^T \begin{bmatrix} \tilde{\Omega} & \tilde{A}^T P F \\ \star & F^T P F \end{bmatrix} \begin{bmatrix} x(k) \\ v(k) \end{bmatrix}. \end{aligned}$$

with  $\mathcal{E}\{\cdot\}$  indicating the mathematical expectation.

When there is no disturbance, we have

$$\Delta V(k) = x^T(k)\tilde{\Omega}x(k). \quad (28)$$

According to (27), one can derive  $\Delta V(k) < 0$  for all  $x(k) \neq 0$ , leading to  $\lim_{k \rightarrow \infty} \mathcal{E}\{\|x(k)\|\} = 0$ . Consequently, system (26) is mean-square stable [30].

When there are some disturbances, one defines the following function.

$$\Phi(k) = \Delta V(k) - \gamma^2 v^T(k)v(k) + \Delta f^T(k)\Delta f(k). \quad (29)$$

Through summing the expectation of each side of (29) from  $k = 0$  to  $k = \infty$ , one arrives

$$\begin{aligned} \tilde{\Phi} &= \mathcal{E}\left\{\sum_{k=0}^{\infty} \Phi(k)\right\} \\ &= \mathcal{E}\{V(\infty)\} - V(0) + \mathcal{E}\left\{\sum_{k=0}^{\infty} \Delta f^T(k)\Delta f(k)\right\} \\ &\quad - \gamma^2 \sum_{k=0}^{\infty} v^T(k)v(k). \end{aligned}$$

Based on the sufficient condition (27), one arrives  $\tilde{\Phi} < 0$ . Furthermore,  $\mathcal{E}\{V(\infty)\} \geq 0$ . As a result, if  $V(0) = 0$ , then  $\|\Delta f\|_{\mathcal{E}_2}^2 < \gamma^2 \|v\|_2^2$  establishes for all nonzero  $v \in l_2[0, \infty)$ . Therefore,  $\Delta f$  satisfies the prescribed  $\mathcal{H}_\infty$  performance index  $\gamma$  [30]. The proof of Theorem 1 is completed. ■

### E. Control Synthesis

The presented condition (27) poses challenges for solution using the LMI toolbox in Matlab, primarily due to two distinct reasons. Firstly, the coupling between the resilient controller gain  $K$  and the Lyapunov matrix  $P$  within (27) renders traditional LMI methods inadequate. Secondly, (27) contains a matrix  $\Delta A(k)$  that is both time-varying and unknown. Consequently, to enable the automated determination of the resilient controller gain  $K$  from (27), we introduce Theorem 2. Theorem 2 is specifically devised to address the aforementioned challenges, decoupling  $K$  and  $P$ , and in the meantime, accommodating the complexities associated with the time-varying and unknown matrix  $\Delta A(k)$ .

*Theorem 2:* Considering the intermittent measurement probability  $\bar{\vartheta}$  and the intermittent transmission probability  $\bar{\varrho}$ , the system (26) is mean-square asymptotically stable with  $\Delta f$  satisfying the predefined  $\mathcal{H}_\infty$  index  $\gamma$ , if there exist matrices  $P \succ 0$ ,  $\bar{P} \succ 0$ , and a scalar  $\varsigma > 0$ , such that

$$\begin{bmatrix} -\bar{P} & 0 & \tilde{B} & 0 & 0 \\ \star & -\bar{P} & A + \bar{\vartheta}\bar{\varrho}BKC & F & N \\ \star & \star & D^T D - P + \varsigma Y^T Y & 0 & 0 \\ \star & \star & \star & -\gamma^2 I & 0 \\ \star & \star & \star & \star & -\varsigma I \end{bmatrix} \prec 0, \quad (30)$$

$$P\bar{P} = I, \quad (31)$$

*Proof:* Utilizing Schur complement, Young's inequality, and Schur complement to (30), one arrives at

$$\begin{bmatrix} \bar{\Omega} + D^T D & \tilde{A}^T \bar{P}^{-1} F \\ \star & F^T Z^{-1} F - \gamma^2 I \end{bmatrix} \prec 0, \quad (32)$$

where  $\bar{\Omega} = \tilde{A}^T \bar{P}^{-1} \tilde{A} + \tilde{B}^T Z \bar{P}^{-1} \tilde{B} - P$ .

Through defining  $\bar{P} = P^{-1}$ , one obtains (27). The proof of Theorem 2 is completed. ■

## IV. RESULTS

### A. System Parameters

Parameters of the multi-energy microgrid and the resilient controller are given as [26]:  $P_W^{\text{ref}} = 2MW$ , corresponding to  $i_{qr} = 1.75A$ , and  $\omega = 1.17\text{rad/s}$ ;  $R_r = 0.00552pu$ ,  $R_s = 0.00491pu$ ,  $L_s = 0.09273pu$ ,  $L_m = 3.96545pu$ ,  $L_r = 0.1pu$ ,  $M_t = 4.5s$ ,  $l = 1s$ ,  $Q = I_2$ ,  $S = I_1$ ,  $\bar{M} = 5$ ,  $K_d = 1$ ,  $K_g = 1$ ,  $K_p = 10$ ,  $K_b = 11$ ,  $T_b = 0.05$ ,  $T_d = 1.2s$ ,  $T_g = 0.1s$ ,  $T_p = 20s$ ,  $R = 0.06$ ,  $N = [0.02; 0.02; 0.02; 0.02]$ ,  $Y = N^T$ ,  $G(k) = 0.1\sin(10k)$ ,  $h = 1s$ ,  $\gamma = 2$ ,  $T_m(k) = 0.02e^{-0.01k}(\text{rand} - 0.5)$  with ‘‘rand’’ meaning a uniform-distributed random number belonging to  $[0, 1]$ ,  $\Delta P_O = 0.05e^{-0.01k}(\text{rand} - 0.5)$ , and  $\Delta P_L = 0.01e^{-0.01k}(\text{rand} - 0.5)$ .

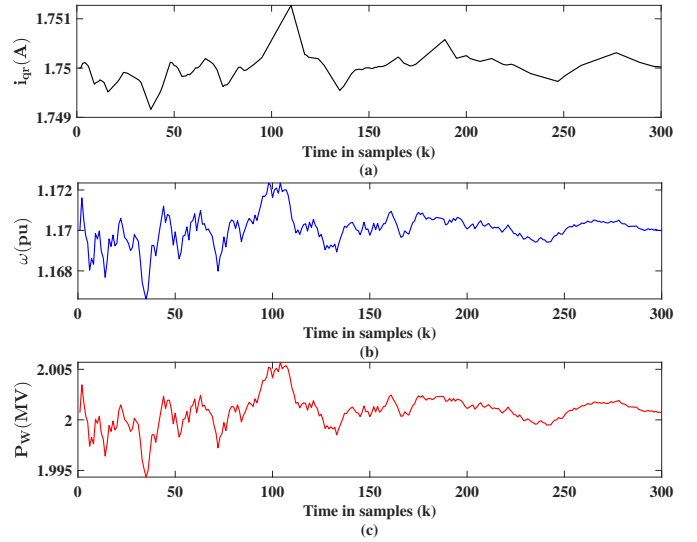


Fig. 3. Dynamics of wind power. (a)  $i_{qr}$ . (b)  $\omega$ . (c)  $P_W$ .

### B. Validations on Rated Wind Power Tracking under MPC

The dynamics of  $i_{qr}$  and  $\omega$  for wind energy generation are depicted in Fig. 3(a)(b), respectively. As seen in Fig. 3(a)(b), the mechanical torque  $T_m$  disturbance causes both  $i_{qr}$  and  $\omega$  to fluctuate around their reference values (1.75A and 1.17rad/s). Fig. 3(c) shows the dynamic of  $P_W$  combined with a substantially greater variation with an amplitude of 0.006MW. Given that  $P_W$  is the product of  $\omega$ ,  $X_3$ , and  $i_{qr}$ , it is reasonable. By simply calculating  $\Delta P_W = P_W - P_W^{\text{ref}}$ , we can obtain the amplitude of wind power fluctuation, which is a kind of disturbance whose impact will be analyzed in Section IV.D.

### C. Validations on Fast LFC with PEVs Participation

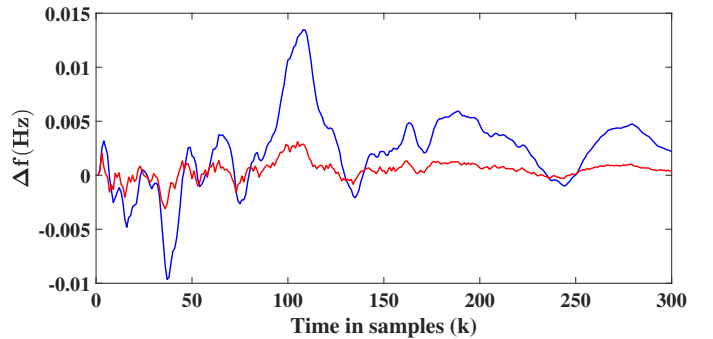


Fig. 4.  $\Delta f$  without and with PEVs. The blue line represents without PEVs. The red line represents with PEVs.

This section presents a comparative assessment of LFC with and without the participation of PEVs, using ideal feedback signals. Fig. 4 illustrates the frequency deviation dynamics for both scenarios. The controller gains for the scenario without the PEV participation are  $[14.4248 \ 0.4492 \ -0.0163]$ , which are automatically determined using the LMI toolbox without considering the PEV in the state

vector. While the gains with the PEV participation are  $[-1.5279 \ 0.3438 \ 0.0372 \ -0.0420]$ , and these gains are automatically selected using the LMI toolbox based on Theorem 2. The results depicted in Fig. 4 clearly demonstrate that the inclusion of PEVs can significantly mitigate the overshoot of the frequency deviation curve compared to the scenario without the PEV participation. This reduction highlights the efficacy of PEVs in enhancing fast frequency regulation. The improvement can be attributed to the rapid charging and discharging capabilities of PEV batteries, which facilitate swift adjustments to frequency deviations.

#### D. Validations on Resilient LFC Considering Intermittent Feedback Signals

To demonstrate the merits of the proposed handling method to intermittent feedback signals, the conventional controller which does not consider the intermittency of the feedback signals is employed as the benchmark. The control gains of the conventional controller are automatically determined using the same methodological framework proposed in the resilient LFC strategy. This determination is executed through the LMI toolbox in Matlab, with the intermittency probability of the feedback signals set to zero.

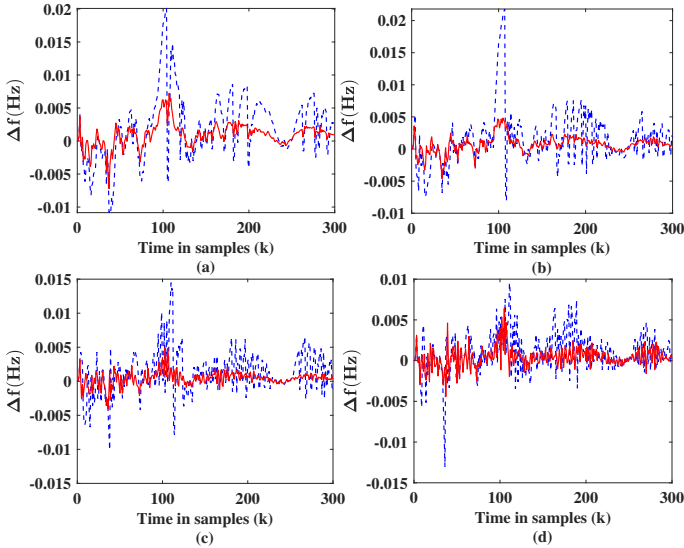


Fig. 5.  $\Delta f$  under different levels of intermittent feedback signals, the blue line indicates the conventional controller while the red line indicates the resilient controller. (a)  $\bar{\nu} = 0.95, \bar{\rho} = 0.95$ . (b)  $\bar{\nu} = 0.95, \bar{\rho} = 0.85$ . (c)  $\bar{\nu} = 0.95, \bar{\rho} = 0.75$ . (d)  $\bar{\nu} = 0.95, \bar{\rho} = 0.65$ .

Fig. 5 compares  $\Delta f$  with different intermittent transmission probabilities  $1 - \bar{\rho} = 0.05, 0.15, 0.25, 0.35$ , while the intermittent measurement probability maintains  $1 - \bar{\nu} = 0.05$ . Table I gives the corresponding resilient control gains, respectively. In Fig. 5, it is obvious that the presented control approach can achieve smaller overshoot and faster convergence than the conventional controller when facing variant power from DRESS with different probabilities of intermittent feedback signals. According to Fig. 5, it is evident that the proposed control approach is capable of achieving superior frequency regulation

dynamics, which verifies the efficacy of the proposed approach under different levels of intermittent feedback signals.

TABLE I  
RESILIENT CONTROL GAINS UNDER  $1 - \bar{\nu} = 0.05$

Intermittent transmission	Resilient controller gains			
$1 - \bar{\rho} = 0.05$	-1.3190	0.2993	-0.0044	-0.0646
$1 - \bar{\rho} = 0.15$	-1.1546	0.2720	-0.0577	-0.0841
$1 - \bar{\rho} = 0.25$	-0.9894	0.2471	-0.0800	-0.0709
$1 - \bar{\rho} = 0.35$	-0.9177	0.2052	-0.1312	-0.0922

Similar to Fig. 5, Fig. 6 compares the frequency deviation dynamics under different intermittent transmission probabilities  $1 - \bar{\rho} = 0.10, 0.20, 0.30, 0.40$ , while the intermittent measurement probability is fixed as  $1 - \bar{\nu} = 0.15$ . Table II gives the corresponding resilient control gains, respectively. The trend observed in Fig. 6 reaffirms the findings from Fig. 5. Specifically, the resilient controller demonstrates superior frequency regulation compared to the traditional controller across different probabilities of intermittent feedback signals. This trend also indicates that the proposed controller is capable of effectively mitigating the impact of load disturbances and wind power imbalance, maintaining better performance despite different probabilities of intermittent feedback signals.

TABLE II  
RESILIENT CONTROLLER GAINS UNDER  $1 - \bar{\nu} = 0.15$

Intermittent transmission	Resilient controller gains			
$1 - \bar{\rho} = 0.10$	-1.0336	0.2628	-0.0492	-0.0684
$1 - \bar{\rho} = 0.20$	-0.9811	0.2376	-0.1017	-0.0805
$1 - \bar{\rho} = 0.30$	-0.8343	0.1842	-0.1306	-0.0769
$1 - \bar{\rho} = 0.40$	-0.5791	0.1955	-0.0606	-0.0370

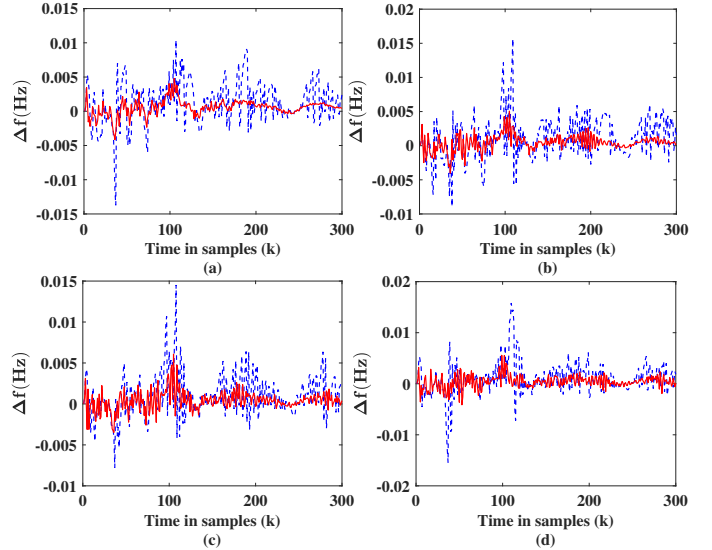


Fig. 6.  $\Delta f$  under different levels of intermittent feedback signals, the blue line indicates the traditional controller while the red line indicates the resilient controller. (a)  $\bar{\nu} = 0.85, \bar{\rho} = 0.90$ . (b)  $\bar{\nu} = 0.85, \bar{\rho} = 0.80$ . (c)  $\bar{\nu} = 0.85, \bar{\rho} = 0.70$ . (d)  $\bar{\nu} = 0.85, \bar{\rho} = 0.60$ .

## V. CONCLUSION

The paper studied the resilient LFC problem considering intermittent feedback signals, including intermittent measurement and intermittent transmission, in the context of multiple



DRESSs. To mitigate the impact of fluctuant power outputs from DRESSs, robust performance index was incorporated in the control gains selection. The sufficient condition was deduced to ensure the mean-square stability for the system, from which the resilient control gains can be automatically determined. The validation results demonstrated that the frequency dynamic can be greatly enhanced with the participation of PEVs, and that the presented resilient frequency controller can well sustain the frequency dynamic under different levels of intermittent feedback signals.

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